Plan of the Lecture

- Review: state-space notions: canonical forms, controllability.
- ► Today's topic: controllability, stability, and pole-zero cancellations; effect of coordinate transformations; conversion of any controllable system to CCF.

Goal: explore the effect of pole-zero cancellations on internal stability; understand the effect of coordinate transformations on the properties of a given state-space model (transfer function; open-loop poles; controllability).

Reading: FPE, Chapter 7

State-Space Realizations

$$u \xrightarrow{\qquad \qquad } \begin{bmatrix} \dot{x} = Ax + Bu \\ y = Cx \end{bmatrix} \xrightarrow{\qquad \qquad } y$$

- ▶ a given transfer function G(s) can be realized using infinitely many state-space models
- certain properties make some realizations preferable to others
- one such property is *controllability*

Controllability Matrix

Consider a single-input system $(u \in \mathbb{R})$:

$$\dot{x} = Ax + Bu, \qquad y = Cx \qquad \qquad x \in \mathbb{R}^n$$

The Controllability Matrix is defined as

$$\mathcal{C}(A,B) = \left[B \mid AB \mid A^2B \mid \dots \mid A^{n-1}B\right]$$

We say that the above system is controllable if its controllability matrix $\mathcal{C}(A, B)$ is *invertible*.

- As we will see later, if the system is controllable, then we may assign arbitrary closed-loop poles by *state feedback* of the form u = -Kx.
- ▶ Whether or not the system is controllable depends on its state-space realization.

Example: Computing $\mathcal{C}(A, B)$

Let's get back to our old friend:

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \underbrace{\begin{pmatrix} 0 & 1 \\ -6 & -5 \end{pmatrix}}_A \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \underbrace{\begin{pmatrix} 0 \\ 1 \end{pmatrix}}_B u, \qquad y = \underbrace{(1 \quad 1)}_C \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

Here, $x \in \mathbb{R}^2 \Longrightarrow A \in \mathbb{R}^{2 \times 2} \Longrightarrow \mathcal{C}(A, B) \in \mathbb{R}^{2 \times 2}$

$$\mathcal{C}(A,B) = \begin{bmatrix} B \mid AB \end{bmatrix} \qquad AB = \begin{pmatrix} 0 & 1 \\ -6 & -5 \end{pmatrix} \begin{pmatrix} 0 \\ 1 \end{pmatrix} = \begin{pmatrix} 1 \\ -5 \end{pmatrix}$$
$$\implies \mathcal{C}(A,B) = \begin{pmatrix} 0 & 1 \\ 1 & -5 \end{pmatrix}$$

Is this system controllable?

$$\det \mathcal{C} = -1 \neq 0 \qquad \Longrightarrow \qquad$$

system is controllable

Controller Canonical Form

A single-input state-space model

$$\dot{x} = Ax + Bu, \qquad y = Cx$$

is said to be in Controller Canonical Form (CCF) is the matrices A, B are of the form

$$A = \begin{pmatrix} 0 & 1 & 0 & \dots & 0 & 0 \\ 0 & 0 & 1 & \dots & 0 & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots & \vdots \\ 0 & 0 & 0 & \dots & 0 & 1 \\ * & * & * & \dots & * & * \end{pmatrix}, \qquad B = \begin{pmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{pmatrix}$$

A system in CCF is always controllable!!

(The proof of this for n > 2 uses the Jordan canonical form, we will not worry about this.)

CCF with Arbitrary Zeros

In our example, we had $G(s) = \frac{s+1}{s^2+5s+6}$, with a minimum-phase zero at z = -1.

Let's consider a general zero location s = z:

$$G(s) = \frac{s - z}{s^2 + 5s + 6}$$

This gives us a CCF realization

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \underbrace{\begin{pmatrix} 0 & 1 \\ -6 & -5 \end{pmatrix}}_A \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \underbrace{\begin{pmatrix} 0 \\ 1 \end{pmatrix}}_B u, \qquad y = \underbrace{(-z \quad 1)}_C \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

Since A, B are the same, $\mathcal{C}(A, B)$ is the same \Longrightarrow the system is still controllable.

A system in CCF is controllable for any locations of the zeros.

OCF with Arbitrary Zeros

Start with the CCF

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \underbrace{\begin{pmatrix} 0 & 1 \\ -6 & -5 \end{pmatrix}}_A \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \underbrace{\begin{pmatrix} 0 \\ 1 \end{pmatrix}}_B u, \qquad y = \underbrace{(-z \quad 1)}_C \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

Convert to OCF: $(A \mapsto A^T, B \mapsto C^T, C \mapsto B^T)$

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \underbrace{\begin{pmatrix} 0 & -6 \\ 1 & -5 \end{pmatrix}}_{\bar{A}=A^T} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \underbrace{\begin{pmatrix} -z \\ 1 \end{pmatrix}}_{\bar{B}=C^T} u, \qquad y = \underbrace{\begin{pmatrix} 0 & 1 \end{pmatrix}}_{\bar{C}=B^T} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

We already know that this system realizes the same t.f. as the original system.

But is it *controllable*?

OCF with Arbitrary Zeros

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \underbrace{\begin{pmatrix} 0 & -6 \\ 1 & -5 \end{pmatrix}}_{\bar{A}=A^T} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \underbrace{\begin{pmatrix} -z \\ 1 \end{pmatrix}}_{\bar{B}=C^T} u, \qquad y = \underbrace{\begin{pmatrix} 0 & 1 \end{pmatrix}}_{\bar{C}=B^T} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

Let's find the controllability matrix:

$$\mathcal{C}(\bar{A},\bar{B}) = \begin{bmatrix} \bar{B} \mid \bar{A}\bar{B} \end{bmatrix} \qquad \bar{A}\bar{B} = \begin{pmatrix} 0 & -6\\ 1 & -5 \end{pmatrix} \begin{pmatrix} -z\\ 1 \end{pmatrix} = \begin{pmatrix} -6\\ -z-5 \end{pmatrix}$$
$$\therefore \ \mathcal{C}(\bar{A},\bar{B}) = \begin{pmatrix} -z & -6\\ 1 & -z-5 \end{pmatrix}$$
$$\det \mathcal{C} = z(z+5) + 6 = z^2 + 5z + 6 = 0 \quad \text{for } z = -2 \text{ or } z = -3$$

The OCF realization of the transfer function $G(s) = \frac{s-z}{s^2+5s+6}$ is not controllable when z = -2 or -3, even though the CCF is always controllable.

Beware of Pole-Zero Cancellations!

The OCF realization of the transfer function

$$G(s) = \frac{s-z}{s^2+5s+6}$$

is not controllable when z = -2 or -3, even though the CCF is always controllable.

Let's examine G(s) when z = -2:

$$G(s) = \frac{s-z}{s^2+5s+6}\Big|_{z=-2} = \frac{s+2}{(s+2)(s+3)} = \frac{1}{s+3}$$

— pole-zero cancellation!

For z = -2, G(s) is a first-order transfer function, which can always be realized by this 1st-order controllable model:

$$\dot{x}_1 = -3x_1 + u, \ y = x_1 \quad \longrightarrow \quad G(s) = \frac{1}{s+3}$$

Beware of Pole-Zero Cancellations!!

We can look at this from another angle: consider the t.f.

$$G(s) = \frac{1}{s+3}$$

We can realize it using a one-dimensional controllable state-space model

$$\dot{x}_1 = -3x_1 + u, \quad y = x_1$$

or a noncontrollable two-dimensional state-space model

$$\begin{pmatrix} \dot{x}_1 \\ \dot{x}_2 \end{pmatrix} = \begin{pmatrix} 0 & -6 \\ 1 & -5 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix} + \begin{pmatrix} 2 \\ 1 \end{pmatrix} u, \qquad y = \begin{pmatrix} 0 & 1 \end{pmatrix} \begin{pmatrix} x_1 \\ x_2 \end{pmatrix}$$

— certainly not the best way to realize a simple t.f.!

Thus, even the *state dimension* of a realization of a given t.f. is not unique!!

Beware of Pole-Zero Cancellations!!

Here is a really bad realization of the t.f.

$$G(s) = \frac{1}{s+3}.$$

Use a two-dimensional model:

$$\begin{aligned} \dot{x}_1 &= -3x_1 + u\\ \dot{x}_2 &= 100x_2\\ y &= x_1 \end{aligned}$$

- ▶ x₂ is not affected by the input u (i.e., it is an uncontrollable mode), and not visible from the output y
- ▶ does not change the transfer function
- ... and yet, horrible to implement: $x_2(t) \propto e^{100t}$

The transfer function can mask undesirable internal state behavior!!

Pole-Zero Cancellations and Stability

- ▶ In case of a pole-zero cancellation, the t.f. contains *much less* information than the state-space model because some dynamics are "hidden."
- These dynamics can be either good (stable) or bad (unstable), but we cannot tell from the t.f.
- Our original definition of stability (no RHP poles) is flawed because there can be RHP eigenvalues of the system matrix A that are canceled by zeros, yet they still have dynamics associated with them.

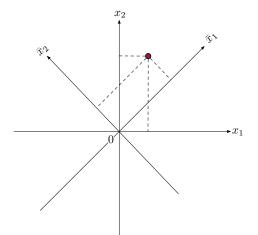
Definition of Internal Stability (State-Space Version): a state-space model with matrices (A, B, C, D) is *internally stable* if all eigenvalues of the A matrix are in LHP.

This is equivalent to having no RHP open-loop poles and no pole-zero cancellations in RHP.

Now that we have seen that a given transfer function can have many different state-space realizations, we would like a systematic procedure of generating such realizations, preferably with favorable properties (like controllability).

One such procedure is by means of *coordinate transformations*.

Coordinate Transformations



 $x \mapsto \bar{x} = Tx,$ $T \in \mathbb{R}^{n \times n}$ nonsingular $x = T^{-1}\bar{x}$ (go back and forth between the coordinate systems)

Coordinate Transformations

For example,

$$\begin{pmatrix} x_1 \\ x_2 \end{pmatrix} \longmapsto \begin{pmatrix} \bar{x}_1 \\ \bar{x}_2 \end{pmatrix} = \begin{pmatrix} x_1 + x_2 \\ x_1 - x_2 \end{pmatrix}$$

This can be represented as

$$\bar{x} = Tx$$
, where $T = \begin{pmatrix} 1 & 1 \\ 1 & -1 \end{pmatrix}$

The transformation is invertible: det T = -2, and

$$T^{-1} = \frac{1}{\det T} \begin{pmatrix} -1 & -1 \\ -1 & 1 \end{pmatrix} = \begin{pmatrix} \frac{1}{2} & \frac{1}{2} \\ \frac{1}{2} & -\frac{1}{2} \end{pmatrix}$$

Or we can see this directly:

$$\bar{x}_1 + \bar{x}_2 = 2x_1; \quad \bar{x}_1 - \bar{x}_2 = 2x_2$$

Coordinate Transformations and State-Space Models Consider a state-space model

$$\dot{x} = Ax + Bu$$
$$y = Cx$$

and a change of coordinates $\bar{x} = Tx$ (*T* invertible). What does the system look like in the new coordinates?

$$\begin{split} \dot{\bar{x}} &= \bar{T}\dot{x} = T\dot{x} & \text{(linearity of derivative)} \\ &= T(Ax + Bu) \\ &= T(AT^{-1}\bar{x} + Bu) & (x = T^{-1}\bar{x}) \\ &= \underbrace{TAT^{-1}}_{\bar{A}}\bar{x} + \underbrace{TB}_{\bar{B}}u \\ y &= Cx \\ &= \underbrace{CT^{-1}}_{\bar{C}}\bar{x} \end{split}$$

$$\dot{x} = Ax + Bu \qquad \xrightarrow{T} \qquad \dot{\bar{x}} = \bar{A}\bar{x} + \bar{B}u$$

 $y = Cx \qquad \qquad y = \bar{C}\bar{x}$

where

$$\bar{A} = TAT^{-1}, \qquad \bar{B} = TB, \qquad \bar{C} = CT^{-1}$$

What happens to

- ▶ the transfer function?
- ▶ the controllability matrix?

$$\begin{aligned} \dot{x} &= Ax + Bu & \xrightarrow{T} & \dot{\bar{x}} &= \bar{A}\bar{x} + \bar{B}u \\ y &= Cx & y &= \bar{C}\bar{x} \end{aligned}$$

where $\bar{A} &= TAT^{-1}$, $\bar{B} &= TB$, $\bar{C} &= CT^{-1}$

Claim: The transfer function doesn't change. Proof:

$$\bar{G}(s) = \bar{C}(Is - \bar{A})^{-1}\bar{B}$$

= $(CT^{-1})(Is - TAT^{-1})^{-1}(TB)$
= $CT^{-1}(TIT^{-1}s - TAT^{-1})^{-1}TB$
= $CT^{-1}[T(Is - A)T^{-1}]^{-1}TB$
= $C\underbrace{T^{-1}T}_{I}(Is - A)^{-1}\underbrace{T^{-1}T}_{I}B$
= $C(Is - A)^{-1}B \equiv G(s)$

$$\begin{aligned} \dot{x} &= Ax + Bu & \xrightarrow{T} & \dot{\bar{x}} &= \bar{A}\bar{x} + \bar{B}u \\ y &= Cx & y &= \bar{C}\bar{x} \end{aligned}$$

where $\bar{A} &= TAT^{-1}$, $\bar{B} &= TB$, $\bar{C} &= CT^{-1}$

The transfer function doesn't change.

In fact:

- open-loop poles don't change
- characteristic polynomial doesn't change:

$$det(Is - \overline{A}) = det(Is - TAT^{-1})$$
$$= det \left[T(Is - A)^{-1}T^{-1}\right]$$
$$= det T \cdot det(Is - A)^{-1} \cdot det T^{-1}$$
$$= det(Is - A)^{-1}$$

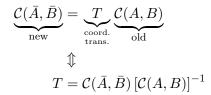
$$\begin{aligned} \dot{x} &= Ax + Bu & \xrightarrow{T} & \dot{\bar{x}} &= \bar{A}\bar{x} + \bar{B}u \\ y &= Cx & y &= \bar{C}\bar{x} \end{aligned}$$
where $\bar{A} &= TAT^{-1}, \quad \bar{B} &= TB, \quad \bar{C} &= CT^{-1} \end{aligned}$

Claim: Controllability doesn't change. Proof: For any k = 0, 1, ..., $\bar{A}^k \bar{B} = (TAT^{-1})^k TB = TA^k T^{-1} TB = TA^k B$ (by induction) Therefore, $C(\bar{A}, \bar{B}) = [TB | TAB | ... | TA^{n-1}B]$ $= T[B | AB | ... | A^{n-1}B]$ = TC(A, B)

Since det $T \neq 0$, det $\mathcal{C}(\bar{A}, \bar{B}) \neq 0$ if and only if det $\mathcal{C}(A, B) \neq 0$. Thus, the new system is controllable if and only if the old one is.

$$\dot{x} = Ax + Bu \qquad \xrightarrow{T} \qquad \dot{\bar{x}} = \bar{A}\bar{x} + \bar{B}u$$
$$y = Cx \qquad \qquad y = \bar{C}\bar{x}$$
where $\bar{A} = TAT^{-1}$, $\bar{B} = TB$, $\bar{C} = CT^{-1}$

Note: The *controllability matrix* does change:



This is a recipe for going from one *controllable* realization of a given t.f. to another.

CCF is the most convenient controllable realization of a given t.f., so we want to *convert a given controllable system to CCF* (useful for control design).

Example: Converting a Controllable System to CCF

Note!! The way I do this is different from the textbook.

Consider
$$A = \begin{pmatrix} -15 & 8 \\ -15 & 7 \end{pmatrix}$$
, $B = \begin{pmatrix} 1 \\ 1 \end{pmatrix}$ (*C* is immaterial).

Convert to CCF if possible.

Step 1: check for controllability.

$$AB = \begin{pmatrix} -15 & 8 \\ -15 & 7 \end{pmatrix} \begin{pmatrix} 1 \\ 1 \end{pmatrix} = \begin{pmatrix} -7 \\ -8 \end{pmatrix} \implies \mathcal{C} = \begin{pmatrix} 1 & -7 \\ 1 & -8 \end{pmatrix}$$
$$\det \mathcal{C} = -1 \qquad - \text{ controllable}$$

Example: Converting a Controllable System to CCF

Step 2: Determine desired $\mathcal{C}(\bar{A}, \bar{B})$.

We need to figure out \overline{A} and \overline{B} .

For CCF, we must have

$$\bar{A} = \begin{pmatrix} 0 & 1 \\ -a_2 & -a_1 \end{pmatrix}, \qquad \bar{B} = \begin{pmatrix} 0 \\ 1 \end{pmatrix},$$

so we need to find the coefficients a_1, a_2 .

Recall: the characteristic polynomial does not change:

$$\det(Is - A) = \det(Is - \bar{A})$$
$$\det\begin{pmatrix} s + 15 & -8\\ 15 & s - 7 \end{pmatrix} = \det\begin{pmatrix} s & -1\\ a_2 & s + a_1 \end{pmatrix}$$
$$(s + 15)(s - 7) + 120 = s(s + a_1) + a_2$$
$$s^2 + 8s + 15 = s^2 + a_1s + a_2$$

Example: Converting a Controllable System to CCF

Step 2: Determine desired $C(\bar{A}, \bar{B})$. We need to figure out \bar{A} and \bar{B} .

For CCF, we must have

$$\bar{A} = \begin{pmatrix} 0 & 1 \\ -a_2 & -a_1 \end{pmatrix}, \qquad \bar{B} = \begin{pmatrix} 0 \\ 1 \end{pmatrix}.$$

We have just computed

$$\bar{A} = \begin{pmatrix} 0 & 1 \\ -15 & -8 \end{pmatrix}, \quad \bar{B} = \begin{pmatrix} 0 \\ 1 \end{pmatrix}$$

Therefore, the new controllability matrix should be

$$\mathcal{C}(\bar{A},\bar{B}) = [\bar{B} \,|\, \bar{A}\bar{B}] = \begin{pmatrix} 0 & 1\\ 1 & -8 \end{pmatrix}$$

Example: Converting a Controllable System to CCF Step 3: Compute T.

Recall: $T = \mathcal{C}(\bar{A}, \bar{B}) \cdot [\mathcal{C}(A, B)]^{-1}$

$$\mathcal{C}(A,B) = \begin{pmatrix} 1 & -7\\ 1 & -8 \end{pmatrix}$$
$$\mathcal{C}(A,B)]^{-1} = \begin{pmatrix} 1 & -7\\ 1 & -8 \end{pmatrix}^{-1}$$
$$= \frac{1}{-1} \begin{pmatrix} -8 & 7\\ -1 & 1 \end{pmatrix} = \begin{pmatrix} 8 & -7\\ 1 & -1 \end{pmatrix}$$
$$\mathcal{C}(\bar{A},\bar{B}) = \begin{pmatrix} 0 & 1\\ 1 & -8 \end{pmatrix}$$
$$T = \begin{pmatrix} 0 & 1\\ 1 & -8 \end{pmatrix} \begin{pmatrix} 8 & -7\\ 1 & -1 \end{pmatrix}$$
$$= \begin{pmatrix} 1 & -1\\ 0 & 1 \end{pmatrix}$$

In the next lecture, we will see why CCF is so useful.