

Plan of the Lecture

- ▶ Review: state-space models of systems; linearization
- ▶ Today's topic: linear systems and their dynamic response

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Reading: FPE, Section 3.1, Appendix A.

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$$y = Cx$$

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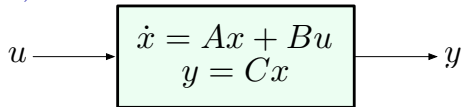
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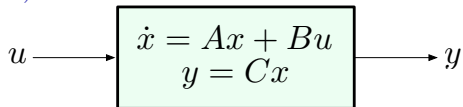
Impulse Response

(Review from ECE 210)



Impulse Response

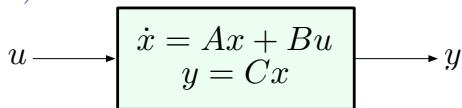
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Unit impulse (or Dirac's δ -function):

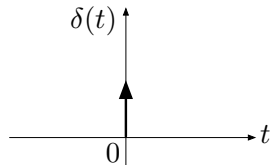
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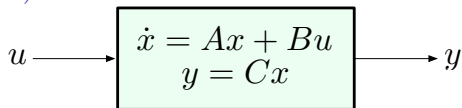
Unit impulse (or Dirac's δ -function):

1. $\delta(t) = 0$ for all $t \neq 0$
2. $\int_{-a}^a \delta(t)dt = 1$ for all $a > 0$



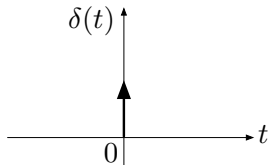
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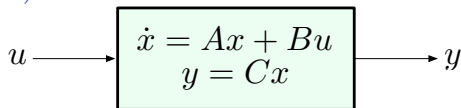
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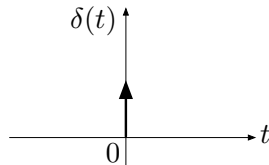
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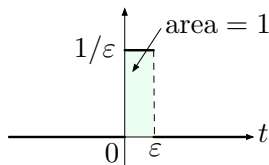


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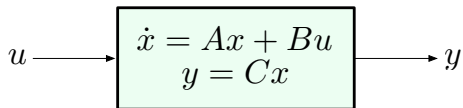


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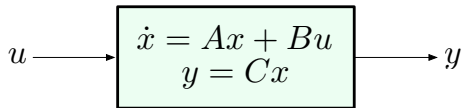
as $\epsilon \rightarrow 0$, the impulse gets taller ($1/\epsilon \rightarrow +\infty$), but the area under its graph remains at 1

Impulse Response



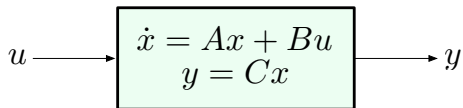
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Impulse Response



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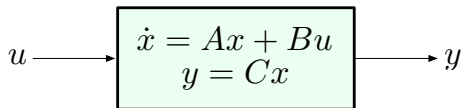


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Consider the input

$$u(t) = \delta(t - \tau) \quad \text{unit impulse applied at } t = \tau$$

Impulse Response



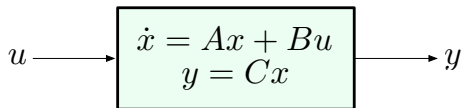
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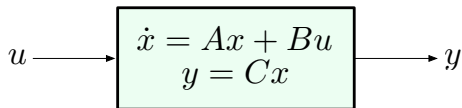
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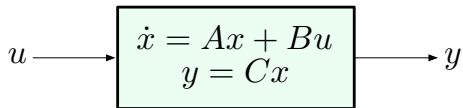
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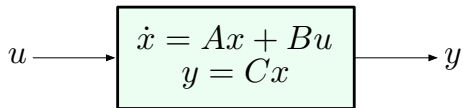
The function h is the *impulse response* of the system.

Impulse Response



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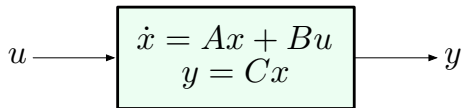
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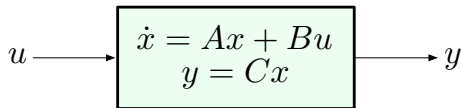
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Questions to consider:

1. If we know h , how can we find the system's response to other (arbitrary) inputs?
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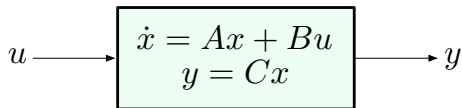
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We will start with Question 1.

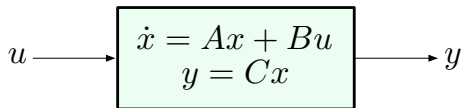
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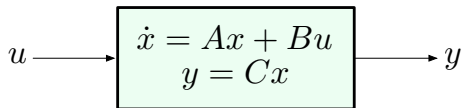
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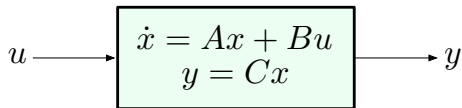
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— any *reasonably regular* function can be represented as an integral of impulses!!

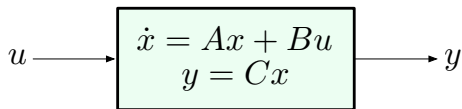
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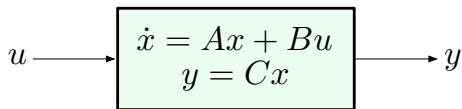
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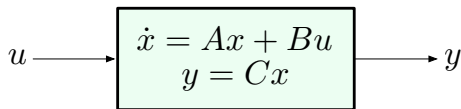
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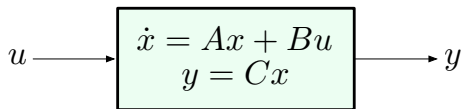
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Impulse Response



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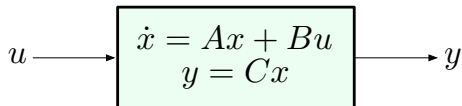
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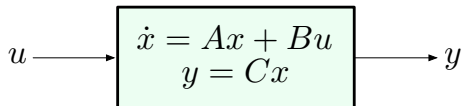
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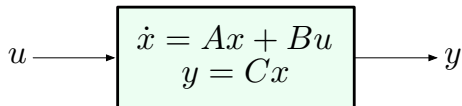
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— the integral that defines $y(t)$ is a

Impulse Response



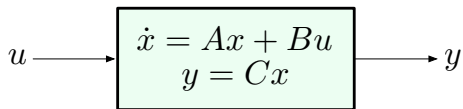
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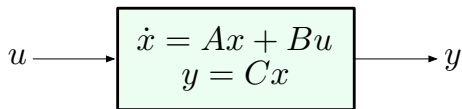
— the integral that defines $y(t)$ is a **convolution** of u and h .

Impulse Response



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Impulse Response

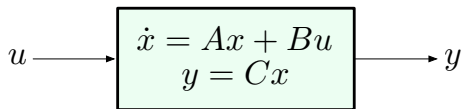


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Conclusion so far: for zero initial conditions, the output is the convolution of the input with the system impulse response:

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Impulse Response



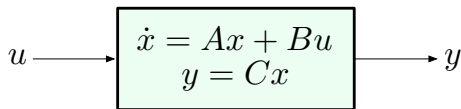
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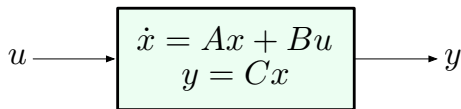
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Laplace Transforms and the Transfer Function

Reminder: the *two-sided* Laplace transform of a function $f(t)$ is

$$F(s) = \int_{-\infty}^{\infty} f(\tau)e^{-s\tau}d\tau, \quad s \in \mathbb{C}$$

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time domain	frequency domain
-------------	------------------

$u(t)$	$U(s)$
--------	--------

$h(t)$	$H(s)$
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$$y(t) \quad Y(s)$$

convolution in time domain \longleftrightarrow multiplication in frequency domain

$$y(t) = h(t) \star u(t) \quad \longleftrightarrow \quad Y(s) = H(s)U(s)$$

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The Laplace transform of the impulse response

$$H(s) = \int_{-\infty}^{\infty} h(\tau)e^{-s\tau}d\tau,$$

is called the **transfer function** of the system.

Laplace Transforms and the Transfer Function

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Laplace Transforms and the Transfer Function

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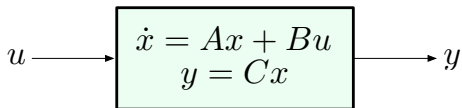
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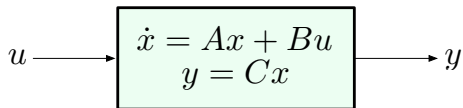


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In this case, we have an **exact formula**:

$$H(s) = C(Is - A)^{-1}B \quad (\text{matrix inversion})$$

$$h(t) = Ce^{At}B, \quad t \geq 0^- \quad (\text{matrix exponential})$$

— will not encounter this until much later in the semester.

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– so, $u(t) = e^{st}$ is multiplied by $H(s)$ to give the output.

Example

$$\dot{y} = -ay + u$$

$$u(t) = e^{st}$$

(think $y = x$, full measurement)

(always assume $u(t) = 0$ for $t < 0$)

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$$H(s) = \frac{1}{s + a} \quad \implies \quad y(t) = \frac{e^{st}}{s + a}$$

Example (continued)

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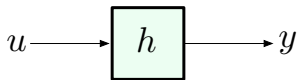
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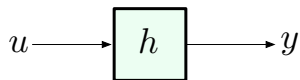
Now we can find the impulse response $h(t)$ by taking the inverse Laplace transform — from tables,

$$h(t) = \begin{cases} e^{-at}, & t \geq 0 \\ 0, & t < 0 \end{cases}$$

Determining the Impulse Response

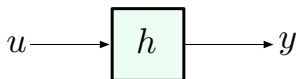


Determining the Impulse Response



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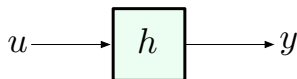


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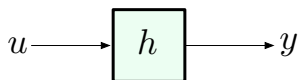
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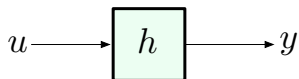
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One idea: inject the input $u(t) = e^{st}$, determine $y(t)$, compute

$$H(s) = \frac{y(t)}{u(t)};$$

repeat for all s of interest.

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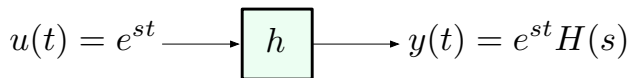
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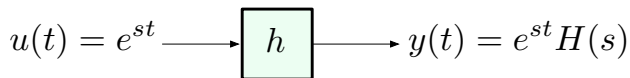
repeat for all s of interest. **Q:** Is this a good idea?

Determining the Impulse Response



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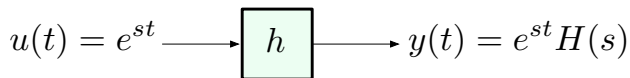
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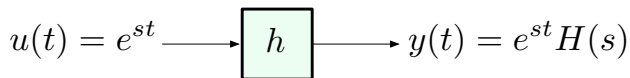


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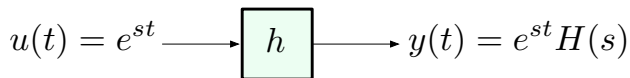
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So we need *sustained, bounded signals* as inputs.

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This is possible if we allow s to take on *complex values*.

Review: Complex Numbers

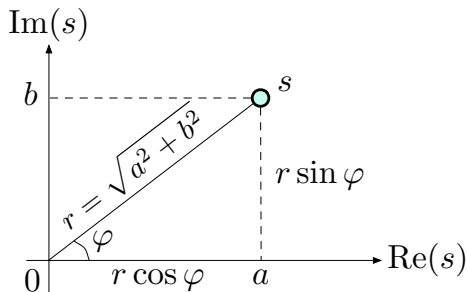
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$$s = r e^{j\varphi}$$

$$r = |s| = \sqrt{a^2 + b^2}$$

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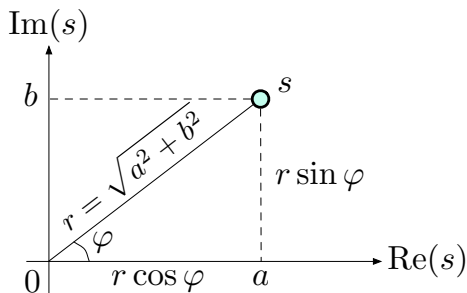
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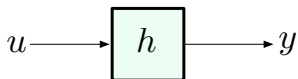
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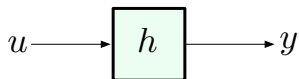
Euler's formula: $e^{j\varphi} = \cos \varphi + j \sin \varphi$

Frequency Response



$$u(t) = A \cos(\omega t) \quad A - \text{amplitude}; \omega - (\text{angular}) \text{ frequency, rad/s}$$

Frequency Response

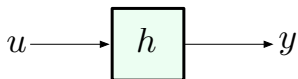


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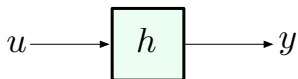
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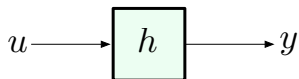
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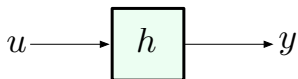
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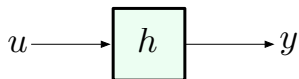
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Frequency Response



$u(t) = A \cos(\omega t)$ A – amplitude; ω – (angular) frequency, rad/s

From Euler's formula:

$$A \cos(\omega t) = \frac{A}{2} (e^{j\omega t} + e^{-j\omega t})$$

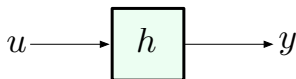
By linearity, the response is

$$y(t) = \frac{A}{2} \left(H(j\omega) e^{j\omega t} + H(-j\omega) e^{-j\omega t} \right)$$

where $H(j\omega) = \int_0^{\infty} h(\tau) e^{-j\omega\tau} d\tau$

$$H(-j\omega) = \int_0^{\infty} \underbrace{h(\tau) e^{j\omega\tau}}_{\text{complex conjugate}} d\tau = \overline{H(j\omega)}$$

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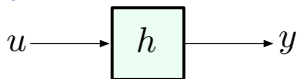
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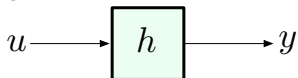
(recall that $h(\tau)$ is real-valued)

Frequency Response



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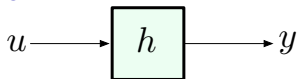
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Frequency Response



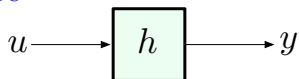
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Frequency Response



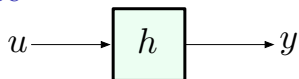
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Frequency Response



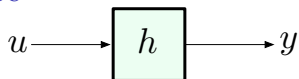
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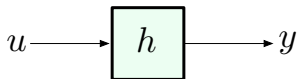
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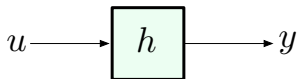
The (steady-state) response to a cosine signal with amplitude A and frequency ω is still a cosine signal with amplitude $AM(\omega)$, same frequency ω , and phase shift $\varphi(\omega)$

Frequency Response



$$u(t) = A \cos(\omega t) \quad \longrightarrow \quad y(t) = A \underbrace{M(\omega)}_{\substack{\text{amplitude} \\ \text{magnification}}} \cos(\omega t + \underbrace{\varphi(\omega)}_{\substack{\text{phase} \\ \text{shift}}})$$

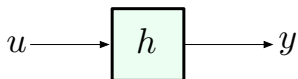
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Still an incomplete picture:

Frequency Response

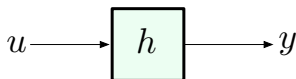


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Still an incomplete picture:

- ▶ What about response to general signals (not necessarily sinusoids)?

Frequency Response

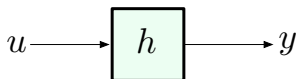


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Still an incomplete picture:

- ▶ What about response to general signals (not necessarily sinusoids)? — always given by $Y(s) = H(s)U(s)$

Frequency Response

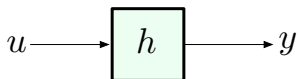


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Frequency Response



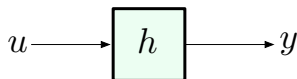
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$$\text{total response} = \begin{array}{l} \text{transient response} \\ \text{(depends on I.C.)} \end{array} + \begin{array}{l} \text{steady-state response} \\ \text{(independent of I.C.)} \end{array}$$

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— need more on Laplace transforms