Topics	Convolution	Freq. Response	DTFT	Summary

Lecture 19: Exam 2 Review

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ECE 401: Signal and Image Analysis, Fall 2021

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1 Topics Covered





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- 3 Frequency Response
- 4 Discrete Time Fourier Transform





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DSP First, chapters 5-7:

- Chapter 5 (MP3; lec. 11-12): FIR Filters
- Chapter 6 (HW3, HW4; lec. 13-14): Frequency Response

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Ochapter 7 (MP4; lec. 15-18): DTFT

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• A system is **linear** if and only if, for any two inputs $x_1[n]$ and $x_2[n]$ that produce outputs $y_1[n]$ and $y_2[n]$,

$$x[n] = x_1[n] + x_2[n] \xrightarrow{\mathcal{H}} y[n] = y_1[n] + y_2[n]$$

 A system is shift-invariant if and only if, for any input x₁[n] that produces output y₁[n],

$$x[n] = x_1[n - n_0] \xrightarrow{\mathcal{H}} y[n] = y_1[n - n_0]$$

• If a system is **linear and shift-invariant** (LSI), then it can be implemented using convolution:

$$y[n] = h[n] * x[n] = \sum_{m} h[m]x[n-m] = \sum_{m} h[n-m]x[m]$$

where h[n] is the impulse response:

$$\delta[n] \xrightarrow{\mathcal{H}} h[n]$$



- A finite impulse response (FIR) filter is one whose impulse response has finite length.
- If *h*[*n*] has finite length, then we can implement the filter using an explicit summation:

$$y[n] = h[n] * x[n] = \sum_{m=0}^{N-1} h[m]x[n-m]$$

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The first difference operator is:

$$x[n] \xrightarrow{\mathcal{H}} y[n] = x[n] - x[n-1]$$

Its impulse response is:

$$h[n] = \delta[n] - \delta[n-1]$$

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The delay operator is:

$$x[n] \stackrel{\mathcal{H}}{\longrightarrow} y[n] = x[n - n_0]$$

Its impulse response is:

$$h[n] = \delta[n - n_0]$$

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The local adder is:

$$x[n] \xrightarrow{\mathcal{H}} y[n] = \sum_{m=0}^{N-1} x[n-m]$$

Its impulse response is:

$$h[n] = egin{cases} 1 & 0 \leq n \leq N-1 \ 0 & ext{otherwise} \end{cases}$$

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 Frequency
 Response

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$$\begin{aligned} x[n] &= e^{j\omega n} \to y[n] = H(\omega)e^{j\omega n} \\ x[n] &= \cos(\omega n) \to y[n] = |H(\omega)|\cos(\omega n + \angle H(\omega)) \\ x[n] &= A\cos(\omega n + \theta) \to y[n] = A|H(\omega)|\cos(\omega n + \theta + \angle H(\omega)) \end{aligned}$$

• where the Frequency Response is given by

$$H(\omega) = \sum_{m} h[m] e^{-j\omega m}$$

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The first difference impulse response is:

$$h[n] = \delta[n] - \delta[n-1]$$

Its frequency response is:

$$egin{aligned} \mathcal{H}(\omega) &= 1 - e^{-j\omega} \ &= e^{j\left(rac{\pi-\omega}{2}
ight)}2\sin(\omega/2) \end{aligned}$$

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The delay operator is:

$$h[n] = \delta[n - n_0]$$

Its frequency response is:

$$H(\omega) = e^{-j\omega n_0}$$

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The delayed local adder's impulse response is:

$$h[n] = egin{cases} 1 & 0 \leq n \leq N-1 \\ 0 & ext{otherwise} \end{cases}$$

Its frequency response is:

$$egin{aligned} \mathcal{H}(\omega) &= \sum_{n=0}^{N-1} e^{-j\omega n} \ &= e^{-rac{j\omega(N-1)}{2}} \mathcal{D}_{N}(\omega), \end{aligned}$$

where $D_N(\omega)$ is the "Dirichlet form," sometimes called the "digital sinc:"

$$D_N(\omega) = rac{\sin(\omega N/2)}{\sin(\omega/2)}$$

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If the input of an LSI system is periodic,

$$x[n] = \sum_{k=-N_0/2}^{(N_0-1)/2} X_k e^{j2\pi kn/N_0}$$

... then the output is

$$y[n] = \sum_{k=-N_0/2}^{(N_0-1)/2} X_k H(k\omega_0) e^{j2\pi kn/N_0}$$

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Cascaded LSI Systems convolve their impulse responses, equivalently, they multiply their frequency responses:

$$y[n] = h[n] * g[n] * x[n] \quad \leftrightarrow \quad Y_k = H(k\omega_0)G(k\omega_0)X_k$$

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Summarv				

The DTFT (discrete time Fourier transform) of any signal is $X(\omega)$, given by

$$X(\omega) = \sum_{n=-\infty}^{\infty} x[n]e^{-j\omega n}$$
$$x[n] = \frac{1}{2\pi} \int_{-\pi}^{\pi} X(\omega)e^{j\omega n}d\omega$$

Particular useful examples include:

$$f[n] = \delta[n] \leftrightarrow F(\omega) = 1$$
$$g[n] = \delta[n - n_0] \leftrightarrow G(\omega) = e^{-j\omega n_0}$$

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Properties worth knowing include:

• Periodicity: $X(\omega + 2\pi) = X(\omega)$

Linearity:

$$z[n] = ax[n] + by[n] \leftrightarrow Z(\omega) = aX(\omega) + bY(\omega)$$

- 2 Time Shift: $x[n n_0] \leftrightarrow e^{-j\omega n_0} X(\omega)$
- Solution Frequency Shift: $e^{j\omega_0 n} x[n] \leftrightarrow X(\omega \omega_0)$
- Iltering is Convolution:

$$y[n] = h[n] * x[n] \leftrightarrow Y(\omega) = H(\omega)X(\omega)$$

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Ideal Filt	ers			

• Ideal Lowpass Filter:

$$H_{LP}(\omega) = \begin{cases} 1 & |\omega| \le \omega_c, \\ 0 & \omega_c < |\omega| \le \pi. \end{cases} \quad \leftrightarrow \quad h_{LP}[m] = \frac{\omega_c}{\pi} \operatorname{sinc}(\omega_c n)$$

• Ideal Bandpass Filter:

$$H_{BP}(\omega) = H_{LP,\omega_2}(\omega) - H_{LP,\omega_1}(\omega)$$

$$\leftrightarrow h_{BP}[n] = \frac{\omega_2}{\pi} \operatorname{sinc}(\omega_2 n) - \frac{\omega_1}{\pi} \operatorname{sinc}(\omega_1 n)$$

• Ideal Highpass Filter:

$$H_{HP}(\omega) = 1 - H_{LP}(\omega) \quad \leftrightarrow \quad h_{HP}[n] = \operatorname{sinc}(\pi n) - \frac{\omega_c}{\pi} \operatorname{sinc}(\omega_c n)$$

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Practica	l Filters			

• Even-symmetric in time (odd length only):

$$h[n] = \begin{cases} h_{\text{ideal}}[n]w[n] & -\frac{N-1}{2} \le n \le \frac{N-1}{2} \\ 0 & \text{otherwise} \end{cases}$$

• Right-sided in time (odd or even length):

$$h[n] = \begin{cases} h_{\text{ideal}} \left[n - \left(\frac{N-1}{2} \right) \right] w[n] & 0 \le n \le N-1 \\ 0 & \text{otherwise} \end{cases}$$

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where w[n] is a finite-length windowing function.

Windows

You need to know these two windowing functions, presented here in their right-sided forms:

• Rectangular Window:

$$w[n] = egin{cases} 1 & 0 \leq n \leq N-1 \ 0 & ext{otherwise} \end{cases} ext{ } \leftrightarrow \quad W(\omega) = e^{-rac{j\omega(N-1)}{2}} D_N(\omega)$$

- Main lobe halfwidth (first null): $\frac{2\pi}{N}$
- First sidelobe level: -11dB
- Hamming Window:

$$w[n] = \begin{cases} 0.54 - 0.46 \cos\left(\frac{2\pi n}{N-1}\right) & 0 \le n \le N-1\\ 0 & \text{otherwise} \end{cases}$$

- Main lobe halfwidth (first null): $\frac{4\pi}{N}$
- First sidelobe level: -44dB

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- Chapter 5: FIR Filters
 - LSI systems, impulse response, convolution
 - first difference, pure delay, local sum
- Ochapter 6: Frequency Response
 - complex exponentials, cosines, periodic signals
 - cascaded systems
 - first difference, pure delay, local sum
- Ohapter 7: DTFT
 - DTFT & frequency response
 - Ideal filters
 - Practical filters; rectangular & Hamming windows

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