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Lecture 13: Frequency Response

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ECE 401: Signal and Image Analysis, Fall 2021

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- When we process a signal, usually, we're trying to enhance the meaningful part, and reduce the noise.
- **Spectrum** helps us to understand which part is meaningful, and which part is noise.
- **Convolution** (a.k.a. filtering) is the tool we use to perform the enhancement.

• Frequency Response of a filter tells us exactly which frequencies it will enhance, and which it will reduce.



• A convolution is exactly the same thing as a **weighted local** average. We give it a special name, because we will use it very often. It's defined as:

$$y[n] = \sum_{m} h[m]f[n-m] = \sum_{m} h[n-m]f[m]$$

• We use the symbol * to mean "convolution:"

$$y[n] = h[n] * f[n] = \sum_{m} h[m]f[n-m] = \sum_{m} h[n-m]f[m]$$

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The **spectrum** of x(t) is the set of frequencies, and their associated phasors,

Spectrum
$$(x(t)) = \{(f_{-N}, a_{-N}), \dots, (f_0, a_0), \dots, (f_N, a_N)\}$$

such that

$$x(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

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One reason the spectrum is useful is that **any** periodic signal can be written as a sum of cosines. Fourier's theorem says that any x(t) that is periodic, i.e.,

$$x(t+T_0)=x(t)$$

can be written as

$$x(t) = \sum_{k=-\infty}^{\infty} X_k e^{j2\pi k F_0 t}$$

which is a special case of the spectrum for periodic signals: $f_k = kF_0$, and $a_k = X_k$, and

$$F_0 = \frac{1}{T_0}$$

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• Fourier Series Analysis (finding the spectrum, given the waveform):

$$X_k = rac{1}{T_0} \int_0^{T_0} x(t) e^{-j2\pi kt/T_0} dt$$

• Fourier Series Synthesis (finding the waveform, given the spectrum):

$$x(t) = \sum_{k=-\infty}^{\infty} X_k e^{j2\pi kt/T_0}$$

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Frequency Response

The **frequency response**, $H(\omega)$, of a filter h[n], is its output in response to a pure tone, expressed as a function of the frequency of the tone.



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• Output of the filter:

$$y[n] = h[n] * x[n]$$
$$= \sum_{m} h[m]x[n-m]$$

• in response to a pure tone:

$$x[n] = e^{j\omega n}$$

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Output of the filter in response to a pure tone:

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$$F[n] = \sum_{m} h[m]x[n-m]$$

= $\sum_{m} h[m]e^{j\omega(n-m)}$
= $e^{j\omega n} \left(\sum_{m} h[m]e^{-j\omega m}\right)$

Notice that the part inside the parentheses is not a function of n. It is not a function of m, because the m gets summed over. It is only a function of ω . It is called the frequency response of the filter, $H(\omega)$.

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Frequency Response

When the input to a filter is a pure tone,

$$x[n]=e^{j\omega n},$$

then its output is the same pure tone, scaled and phase shifted by a complex number called the **frequency response** $H(\omega)$:

$$y[n] = H(\omega)e^{j\omega n}$$

The frequency response is related to the impulse response as

$$H(\omega) = \sum_{m} h[m] e^{-j\omega m}$$

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Remember that taking the difference between samples can be written as a convolution:

$$y[n] = x[n] - x[n-1] = h[n] * x[n],$$

where

$$h[n] = egin{cases} 1 & n=0 \ -1 & n=1 \ 0 & ext{otherwise} \end{cases}$$

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Review Frequency Response Example Superposition Example Example Summary on 00000 00000000 00000000 000000000 000000000 000000000 000000000 Example: First Difference

Suppose that the input is a pure tone:

$$x[n] = e^{j\omega n}$$

Then the output will be

$$y[n] = x[n] - x[n-1]$$

= $e^{j\omega n} - e^{j\omega(n-1)}$
= $H(\omega)e^{j\omega n}$,

where

$$H(\omega) = 1 - e^{-j\omega}$$

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So we have some pure-tone input,

$$x[n] = e^{j\omega n}$$

... and we send it through a first-difference system:

$$y[n] = x[n] - x[n-1]$$

... and what we get, at the output, is a pure tone, scaled by $|H(\omega)|$, and phase-shifted by $\angle H(\omega)$:

$$y[n] = H(\omega)e^{j\omega n}$$

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- How much is the scaling?
- How much is the phase shift?

Let's find out.

$$H(\omega) = 1 - e^{-j\omega}$$

= $e^{-j\frac{\omega}{2}} \left(e^{j\frac{\omega}{2}} - e^{-j\frac{\omega}{2}} \right)$
= $e^{-j\frac{\omega}{2}} \left(2j\sin\left(\frac{\omega}{2}\right) \right)$
= $\left(2\sin\left(\frac{\omega}{2}\right) \right) \left(e^{j\left(\frac{\pi-\omega}{2}\right)} \right)$

So the magnitude and phase response are:

$$|H(\omega)| = 2\sin\left(\frac{\omega}{2}\right)$$

 $\angle H(\omega) = \frac{\pi - \omega}{2}$

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First Difference: Magnitude Response

$$|H(\omega)| = 2\sin\left(\frac{\omega}{2}\right)$$



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First Difference Filter at $\omega = 0$



Suppose we put in the signal $x[n] = ej\omega n$, but at the frequency $\omega = 0$. At that frequency, x[n] = 1. So

$$y[n] = x[n] - x[n-1] = 0$$

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First Difference Filter at $\omega = \pi$



Frequency $\omega = \pi$ means the input is $(-1)^n$:

$$x[n]=e^{j\pi n}=(-1)^n=egin{cases}1&n ext{ is even}\-1&n ext{ is odd}\end{cases}$$

So

$$y[n] = x[n] - x[n-1] = 2x[n]$$

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Frequency $\omega = \frac{\pi}{2}$ means the input is j^n :

$$x[n] = e^{j\frac{\pi n}{2}} = j^n$$

The frequency response is:

$$G\left(\frac{\pi}{2}\right) = 1 - e^{-j\frac{\pi}{2}} = 1 - \left(\frac{1}{j}\right),$$

The output is

$$y[n] = x[n] - x[n-1] = j^n - j^{n-1} = \left(1 - \frac{1}{j}\right)j^n$$

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Superposition and the Frequency Response

The frequency response obeys the principle of **superposition**, meaning that if the input is the sum of two pure tones:

$$x[n] = e^{j\omega_1 n} + e^{j\omega_2 n},$$

then the output is the sum of the same two tones, each scaled by the corresponding frequency response:

$$y[n] = H(\omega_1)e^{j\omega_1 n} + H(\omega_2)e^{j\omega_2 n}$$

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There are no complex exponentials in the real world. Instead, we'd like to know the output in response to a cosine input. Fortunately, a cosine is the sum of two complex exponentials:

$$x[n] = \cos(\omega n) = \frac{1}{2}e^{j\omega n} + \frac{1}{2}e^{-j\omega n},$$

therefore,

$$y[n] = \frac{H(\omega)}{2}e^{j\omega n} + \frac{H(-\omega)}{2}e^{-j\omega n}$$

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What is $H(-\omega)$? Remember the definition:

$$H(\omega) = \sum_m h[m] e^{-j\omega m}$$

Replacing every ω with a $-\omega$ gives:

$$H(-\omega)=\sum_m h[m]e^{j\omega m}.$$

Notice that h[m] is real-valued, so the only complex number on the RHS is $e^{j\omega m}$. But

$$e^{j\omega m} = \left(e^{-j\omega m}
ight)^*$$

so

$$H(-\omega) = H^*(\omega)$$

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$$y[n] = \frac{H(\omega)}{2}e^{j\omega n} + \frac{H^*(\omega)}{2}e^{-j\omega n}$$
$$= \frac{|H(\omega)|}{2}e^{j\angle H(\omega)}e^{j\omega n} + \frac{|H(\omega)|}{2}e^{-j\angle H(\omega)}e^{-j\omega n}$$
$$= |H(\omega)|\cos(\omega n + \angle H(\omega))$$

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Response to a Cosine

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 $x[n] = \cos(\omega n)$

...then ...

$$y[n] = |H(\omega)| \cos(\omega n + \angle H(\omega))$$

Magnitude and Phase Responses

- The Magnitude Response |H(ω)| tells you by how much a pure tone at ω will be scaled.
- The Phase Response ∠H(ω) tells you by how much a pure tone at ω will be advanced in phase.

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Remember that the first difference, y[n] = x[n] - x[n-1], is supposed to sort of approximate a derivative operator:

$$y(t) \approx rac{d}{dt} x(t)$$

If the input is a cosine, what is the output?

$$\frac{d}{dt}\cos\left(\omega t\right) = -\omega\sin\left(\omega t\right) = \omega\cos\left(\omega t + \frac{\pi}{2}\right)$$

Does the first-difference operator behave the same way (multiply by a magnitude of $|H(\omega)| = \omega$, phase shift by $+\frac{\pi}{2}$ radians so that cosine turns into negative sine)?

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 Example: First Difference

Frequency response of the first difference filter is

$${\it H}(\omega)=1-e^{-j\omega}$$

Let's try to convert it to polar form, so we can find its magnitude and phase:

$$egin{aligned} \mathcal{H}(\omega) &= e^{-jrac{\omega}{2}} \left(e^{jrac{\omega}{2}} - e^{-jrac{\omega}{2}}
ight) \ &= e^{-jrac{\omega}{2}} \left(2j\sin\left(rac{\omega}{2}
ight)
ight) \ &= \left(2\sin\left(rac{\omega}{2}
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ight) \left(e^{j\left(rac{\pi-\omega}{2}
ight)}
ight) \end{aligned}$$

So the magnitude and phase response are:

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$$|H(\omega)| = 2\sin\left(\frac{\omega}{2}\right)$$

 $\angle H(\omega) = \frac{\pi - \omega}{2}$

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Taking the derivative of a cosine scales it by ω . The first-difference filter scales it by $|H(\omega)| = 2\sin(\omega/2)$, which is almost the same, but not quite:



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Taking the derivative of a cosine shifts it, in phase, by $+\frac{\pi}{2}$ radians, so that the cosine turns into a negative sine. The first-difference filter shifts it by $\angle H(\omega) = \frac{\pi - \omega}{2}$, which is the same at very low frequencies, but very different at high frequencies.



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Putting it all together, if the input is $x[n] = cos(\omega n)$, the output is

$$y[n] = |H(\omega)| \cos(\omega n + \angle H(\omega)) = 2\sin\left(\frac{\omega}{2}\right) \cos\left(\omega n + \frac{\pi - \omega}{2}\right)$$

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$$y[n] = 2\sin\left(\frac{\omega}{2}\right)\cos\left(\omega n + \frac{\pi - \omega}{2}\right)$$

At very low frequencies, the output is almost $-\sin(\omega n)$, but with very low amplitude:



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$$y[n] = 2\sin\left(\frac{\omega}{2}\right)\cos\left(\omega n + \frac{\pi - \omega}{2}\right)$$

At intermediate frequencies, the phase shift between the input and output is reduced:



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$$y[n] = 2\sin\left(\frac{\omega}{2}\right)\cos\left(\omega n + \frac{\pi - \omega}{2}\right)$$

At very high frequencies, the phase shift between input and output is eliminated – the output is a cosine, just like the input:



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Consider the following system:

$$y[n] = x[n-n_0]$$

This can be written as a convolution, with impulse response

$$h[n] = \delta[n - n_0]$$

What is $H(\omega)$? Using $H(\omega)$, can you find the output of this system if $x[n] = \cos(\omega n)$?

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 $\bullet~\mbox{Tones}~\mbox{in}~\rightarrow~\mbox{Tones}~\mbox{out}$

$$\begin{aligned} x[n] &= e^{j\omega n} \to y[n] = H(\omega)e^{j\omega n} \\ x[n] &= \cos(\omega n) \to y[n] = |H(\omega)|\cos(\omega n + \angle H(\omega)) \\ x[n] &= A\cos(\omega n + \theta) \to y[n] = A|H(\omega)|\cos(\omega n + \theta + \angle H(\omega)) \end{aligned}$$

• where the Frequency Response is given by

$$H(\omega) = \sum_{m} h[m] e^{-j\omega m}$$

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