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## Lecture 9: Exam 1 Review

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ECE 401: Signal and Image Analysis, Fall 2021

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- HW1: Phasors
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Phasors					

$$\begin{aligned} x(t) &= A\cos\left(2\pi ft + \theta\right) \\ &= \Re\left\{ze^{j2\pi ft}\right\} \\ &= \frac{1}{2}z^*e^{-j2\pi ft} + \frac{1}{2}ze^{j2\pi ft} \end{aligned}$$

where

$$z = Ae^{j\theta}$$



#### How do you add

$$z(t) = A\cos(2\pi ft + \theta) + B\cos(2\pi ft + \phi)?$$

#### Answer:

$$z = (A\cos\theta + B\cos\phi) + j(A\sin\theta + B\sin\phi)$$
$$z(t) = \Re\left\{ze^{j2\pi ft}\right\}$$

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The **spectrum** of x(t) is the set of frequencies, and their associated phasors,

Spectrum 
$$(x(t)) = \{(f_{-N}, a_{-N}), \dots, (f_0, a_0), \dots, (f_N, a_N)\}$$

such that

$$x(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

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The **spectrum plot** of a periodic signal is a plot with

- frequency on the X-axis,
- showing a vertical spike at each frequency component,
- each of which is labeled with the corresponding phasor.

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 Property #1: Scaling

Suppose we have a signal

$$\mathbf{x}(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

Suppose we scale it by a factor of G:

$$y(t) = Gx(t)$$

That just means that we scale each of the coefficients by G:

$$y(t) = \sum_{k=-N}^{N} (Ga_k) e^{j2\pi f_k t}$$

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# Topics Phasors Spectrum Fourier Sampling Summary 00 000 00000 00000 00000 00 Property #2: Adding a constant

Suppose we have a signal

$$x(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

Suppose we add a constant, C:

$$y(t) = x(t) + C$$

That just means that we add that constant to  $a_0$ :

$$y(t) = (a_0 + C) + \sum_{k \neq 0} a_k e^{j2\pi f_k t}$$

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Suppose we have two signals:

$$\begin{aligned} x(t) &= \sum_{n=-N}^{N} a'_n e^{j2\pi f'_n t} \\ y(t) &= \sum_{m=-M}^{M} a''_m e^{j2\pi f''_m t} \end{aligned}$$

and we add them together:

$$z(t) = x(t) + y(t) = \sum_{k} a_k e^{j2\pi f_k t}$$

where, if a frequency  $f_k$  comes from both x(t) and y(t), then we have to do phasor addition:

If 
$$f_k = f'_n = f''_m$$
 then  $a_k = a'_n + a''_m$ 

Suppose we have a signal

$$x(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

and we want to time shift it by  $\tau$  seconds:

$$y(t) = x(t-\tau)$$

Time shift corresponds to a **phase shift** of each spectral component:

$$y(t) = \sum_{k=-N}^{N} \left(a_k e^{-j2\pi f_k \tau}\right) e^{j2\pi f_k t}$$

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Suppose we have a signal

$$x(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

and we want to shift it in frequency by some constant overall shift, F:

$$y(t) = \sum_{k=-N}^{N} a_k e^{j2\pi(f_k+F)t}$$

Frequency shift corresponds to amplitude modulation (multiplying it by a complex exponential at the carrier frequency F):

$$y(t) = x(t)e^{j2\pi Ft}$$

Suppose we have a signal

$$x(t) = \sum_{k=-N}^{N} a_k e^{j2\pi f_k t}$$

and we want to differentiate it:

$$y(t) = \frac{dx}{dt}$$

Differentiation corresponds to scaling each spectral coefficient by  $j2\pi f_k$ :

$$y(t) = \sum_{k=-N}^{N} (j2\pi f_k a_k) e^{j2\pi f_k t}$$

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• Analysis (finding the spectrum, given the signal):

$$X_k = rac{1}{T_0} \int_0^{T_0} x(t) e^{-j2\pi kt/T_0} dt$$

• Synthesis (finding the signal, given the spectrum):

$$x(t) = \sum_{k=-\infty}^{\infty} X_k e^{j2\pi kt/T_0}$$

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• Analysis (finding the spectrum, given the signal):

$$X_{k} = \frac{1}{N_{0}} \sum_{0}^{N_{0}-1} x[n] e^{-j2\pi kn/N_{0}}$$

• Synthesis (finding the signal, given the spectrum):

$$x[n] = \sum_{k} X_k e^{j2\pi kn/N_0}$$

where the sum is over any set of  $N_0$  consecutive harmonics.

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• Scaling:

$$y(t) = Gx(t) \Leftrightarrow Y_k = GX_k$$

Add a Constant:

$$y(t) = x(t) + C \Leftrightarrow Y_k = \begin{cases} X_0 + C & k = 0 \\ X_k & \text{otherwise} \end{cases}$$

• Add Signals: Suppose that x(t) and y(t) have the same fundamental frequency, then

$$z(t) = x(t) + y(t) \Leftrightarrow Z_k = X_k + Y_k$$

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• Time Shift: Shifting to the right, in time, by  $\tau$  seconds:

$$y(t) = x(t - \tau) \Leftrightarrow Y_k = a_k e^{-j2\pi k F_0 \tau}$$

• Frequency Shift: Shifting upward in frequency by F Hertz:

$$y(t) = x(t)e^{j2\pi dF_0 t} \Leftrightarrow Y_k = X_{k-d}$$

• Differentiation:

$$y(t) = \frac{dx}{dt} \Leftrightarrow Y_k = j2\pi k F_0 X_k$$

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# Topics Phasors Spectrum Fourier Sampling Summary 00 000 00000 00000 00000 00 How to sample a continuous-time signal

Suppose you have some continuous-time signal, x(t), and you'd like to sample it, in order to store the sample values in a computer. The samples are collected once every  $T_s = \frac{1}{E_s}$  seconds:

$$x[n] = x(t = nT_s)$$

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The spectrum plot of a **discrete-time periodic signal** is a regular spectrum plot, but with the X-axis relabeled. Instead of frequency in Hertz=  $\begin{bmatrix} cycles \\ second \end{bmatrix}$ , we use

$$\omega \left[ \frac{\text{radians}}{\text{sample}} \right] = \frac{2\pi \left[ \frac{\text{radians}}{\text{cycle}} \right] f \left[ \frac{\text{cycles}}{\text{second}} \right]}{F_s \left[ \frac{\text{samples}}{\text{second}} \right]}$$

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Aliasing					

- A sampled sinusoid can be reconstructed perfectly if the Nyquist criterion is met,  $f < \frac{F_s}{2}$ .
- If the Nyquist criterion is violated, then:
  - If  $\frac{F_s}{2} < f < F_s$ , then it will be aliased to

$$f_a = F_s - f$$
$$z_a = z^*$$

i.e., the sign of all sines will be reversed. • If  $F_s < f < \frac{3F_s}{2}$ , then it will be aliased to

$$f_a = f - F_s$$
$$z_a = z$$





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Interpol	ation				

Interpolation is the general method for reconstructing a continuous-time signal from its samples. The formula is:

$$y(t) = \sum_{n=-\infty}^{\infty} y[n]p(t-nT_s)$$

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Interpo	lation ke	ernels			
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- Piece-wise constant interpolation = interpolate using a rectangle
- Piece-wise linear interpolation = interpolate using a triangle

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• Ideal interpolation = interpolate using a sinc



For example, suppose that the pulse is just a rectangle,

$$p(t) = egin{cases} 1 & -rac{T_S}{2} \leq t < rac{T_S}{2} \ 0 & ext{otherwise} \end{cases}$$





The result is a piece-wise constant interpolation of the digital signal:



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The rectangular pulse has the disadvantage that y(t) is discontinuous. We can eliminate the discontinuities by using a triangular pulse:

$$p(t) = egin{cases} 1 - rac{|t|}{T_{\mathcal{S}}} & -T_{\mathcal{S}} \leq t < T_{\mathcal{S}} \ 0 & ext{otherwise} \end{cases}$$



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### The result is a piece-wise linear interpolation of the digital signal:



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If a signal has all its energy at frequencies below Nyquist  $(f < \frac{F_s}{2})$ , then it can be perfectly reconstructed using sinc interpolation:

$$p(t) = rac{\sin(\pi t/T_S)}{\pi t/T_S}$$



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# Topics Phasors Spectrum Fourier Sampling Summary Sinc pulse = ideal bandlimited interpolation

If a signal has all its energy at frequencies below Nyquist  $(f < \frac{F_s}{2})$ , then it can be perfectly reconstructed using sinc interpolation:



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