From Wikipedia: [...] a relaxation oscillator is a nonlinear electronic oscillator that produces a non-sinusoidal output signal, such as a triangle or a square wave.

Our low-power oscillator from Lab 5 of ECE110 uses the nonlinear Schmitt trigger (inverter) to produce both a triangular waveform (at the Schmitt trigger input) and a square waveform (at the Schmitt trigger output).

**Prerequisites**

- Laboratory Exercise #5, the construction of a simple low-power oscillator circuit.
- Familiarity with diode circuits.

**Parts Needed**

- None. This is a reading exercise with some limited exposure to algebra and calculus.
At Home: This entire exercise may be completed anywhere!

Schmitt Trigger Parameters
To understand the operation of the oscillator, we first need to understand the operation of the Schmitt trigger inverter. Find the datasheet for the CD40106 Schmitt Trigger Inverter. The datasheet will describe a hysteresis (a form of memory) within the device where the input/output relationship for changing input values will depend on the time history of the input. For example, if the input voltage $V_{IN}$ starts at 0 volts (ground) and climbs, the output voltage $V_{OUT}$ will remain high until the input voltage reaches the value $V_p$ as demonstrated in Figure 2. At this point, the output voltage will drop to 0 volts. As the input voltage then falls back below $V_p$, the output voltage persists in staying low (0 volts) until finally the input falls below a value of $V_n$. This means that there is not a one-to-one relationship between $V_{IN}$ and $V_{OUT}$ like we are mostly accustomed to in previous math courses. This relationship is graphed in Figure 2. We consider $V_p$ to be the positive-going threshold voltage and $V_n$ to be the negative-going threshold voltage of the Schmitt Trigger.

![Figure 2: The input/output relationship of the Schmitt trigger from Texas Instruments, the TI 40106.](image)

Question 1: Use the datasheet to argue that the positive-going threshold voltage may be given by $V_p \approx 0.4 \ V_{DD}$ for power supply voltages of $V_{DD}$ between 5 and 10 volts.

Question 2: Use the datasheet to argue that the negative-going threshold voltage may be given by $V_n \approx 0.5 \ V_{DD}$ for power supply voltages of $V_{DD}$ between 5 and 10 volts.

If you are curious, you can set up your own laboratory experiment to improve the approximations $V_p \approx 0.5 \ V_{DD}$ and $V_n \approx 0.4 \ V_{DD}$.
Capacitor Charging and Discharging

When a capacitor is charged by a constant (DC) voltage supply of $V_{DD}$, the time-domain voltage across the capacitor is given as

$$V_1(t) = (V_i - V_{DD})\left(1 - e^{-\frac{t}{RC}} \right) + V_{DD}$$

where $C$ is the capacitance being charged, $V_i$ is the initial voltage on the capacitor at time $t = 0$, and $R$ is the series resistance within the charging path. We say that the voltage is asymptotically approaching $V_{DD}$, although most of the charging occurs in a short time span on the order of the product $R \times C$ (often called the time constant). In Figure 3, initial voltage $V_i = 0$ volts.

![Figure 3: The waveform $V_1$ across a capacitor while charging to $V_{DD}$.](image)

When a capacitor is being discharged to ground voltage (0 V), the time-domain voltage across the capacitor is given by

$$V_1(t) = V_{start}e^{-\frac{t}{RC}}$$

where we are assuming the voltage across the capacitor is $V_{start}$ at the beginning of the discharge. As before, the decay of the capacitor to 0 volts follows an asymptotic path with much of the decay occurring in the order of magnitude of $RC$. 

**Equation Reference:** ECE210 textbook, page 97, Analogy Signals and Systems by Kudeki and Munson.
With this understanding, let’s investigate what happens during the charging and discharging cycles of the capacitor within our oscillator.

The Relaxation Oscillator

The voltage across the capacitor in our working oscillator charges and discharges as would be expected with any capacitor. The difference lies in the fact that the capacitor is never allowed to fully charge or discharge due to a circuit feedback path.

To explain why the capacitor never completely charges or discharges, we need to use a model for the inverter that allows us to simplify the circuit into something we are more familiar with. First, we make use of the fact that the input to the Schmitt trigger draws very little current. In fact, the current flowing into the Schmitt trigger is so small, we model it as an open circuit! This is true whether we are charging or discharging the capacitor.
For the output of the Schmitt trigger, we will need two models. 1) When the input voltage $V_1$ is small, the output of the Schmitt trigger output is high (near $V_{DD}$). Therefore, for the charging cycle, the oscillator circuit can be modeled by Figure 6a. 2) When the input voltage is high, the Schmitt trigger output is high (near ground voltage, 0 V) and the oscillator circuit can be modelled by Figure 6b.

During the charging cycle, the capacitor will asymptotically charging towards $V_{DD}$. Of course, it will stop charging at the point that $V_1(t_2) = V_p$ (see Figure 7). We can compute the time required for the charge time $\Delta t = t_2 - t_1$, by substituting the correct parameters and solving the earlier equation for $t_2$.

$$V_p = (V_n - V_{DD}) \left(1 - e^{-\frac{\Delta t}{RC}}\right) + V_{DD}$$

$$\frac{V_p - V_{DD}}{V_n - V_{DD}} = e^{-\frac{\Delta t}{RC}}$$

$$\ln \left(\frac{V_p - V_{DD}}{V_n - V_{DD}}\right) = -\frac{\Delta t}{RC}$$

$\ln(x)$ is the “natural log” of $x$. We make use of the fact that $\ln(x) - \ln(y) = \ln\left(\frac{x}{y}\right)$.
\[ t_{\text{charge}} = \Delta t = RC \ln \left( \frac{V_p - V_{DD}}{V_n - V_{DD}} \right) = RC \ln \left( \frac{V_{DD} - V_p}{V_{DD} - V_n} \right) \]

\[ t_{\text{discharge}} = t_3 = RC \ln \left( \frac{V_p}{V_n} \right) \]

Since the capacitor must both charge and discharge in every period of the oscillatory waveform, the period is given by

\[ T = t_{\text{charge}} + t_{\text{discharge}} = RC \ln \left( \frac{(V_{DD} - V_n)V_p}{(V_{DD} - V_p)V_n} \right) = \ln \left( \frac{(1 - 0.4)0.5}{(1 - 0.5)0.4} \right) RC = \ln \left( \frac{0.6}{0.4} \right) RC = 0.41 RC \]

and the frequency of oscillation is given by

\[ f = \frac{1}{T} = \frac{1}{0.41 RC} \]

**Figure 7:** Focus on the charging and discharging intervals for a Schmitt-trigger-based oscillator.

The time required to discharge from \( V_p \) to \( V_n \) can be computed from \( V_n = V_p e^{-\frac{t_3}{RC}} \):

\[ t_{\text{discharge}} = t_3 = RC \ln \left( \frac{V_p}{V_n} \right) \]

Question 3: What happens to the frequency of oscillation when the capacitance is doubled? What happens to the frequency of oscillation when the resistance is doubled?
**Drawing Conclusions**

**Question 4:** *Think about it...* This analysis changes in Lab 8 when diodes are placed in the feedback and feedforward paths. Redraw the charging and discharging circuit schematics (similar to Figure 6) including a diode in the path for each. How do you think these diodes will affect $t_{charge}$ and $t_{discharge}$?

**Learning Objectives**

- To apply capacitor charging and discharging formulas to the relaxation oscillator and estimate oscillatory frequency.
- To use circuit models to reduce a challenging problem to one that can be solved using basic circuit analysis.

**Explore Even More!**

The oscillator is a key component in *Explore More! Module: The Voltage Comparator* where we learn that the voltage $V_1$ must be protected by a “buffer” in order to be utilized in a Pulse-Width Modulation waveform generator. Eventually, the oscillator is a key component in the *Explore More! Module: PWM Control via an Active Sensor.*